

# CASCADABLE BROADBAND GaAs MMIC AMPLIFIER DC TO 10GHz

RoHS Compliant & Pb-Free Product Package Style: Micro-X, 4-Pin, Plastic

**NLB-300** 

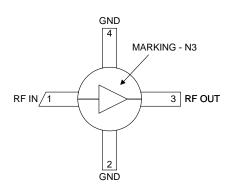


#### **Features**

- Reliable, Low-Cost HBT Design
- 13.0dB Gain, +11.1dBm P1dB@2GHz
- High P1dB of +14.1dBm@6.0GHz and +12.7dBm@10.0GHz
- Single Power Supply Operation
- 50Ω I/O Matched for High Freq. Use

## **Applications**

- Narrow and Broadband Commercial and Military Radio Designs
- Linear and Saturated Amplifiers
- Gain Stage or Driver Amplifiers for MW Radio/Optical Designs (PTP/PMP/ LMDS/UNII/VSAT/WLAN/Cellular/DWDM)



Functional Block Diagram

## **Product Description**

The NLB-300 cascadable broadband InGaP/GaAs MMIC amplifier is a low-cost, high-performance solution for general purpose RF and microwave amplification needs. This  $50\Omega$  gain block is based on a reliable HBT proprietary MMIC design, providing unsurpassed performance for small-signal applications. Designed with an external bias resistor, the NLB-300 provides flexibility and stability. The NLB-300 is packaged in a low-cost, surface-mount plastic package, providing ease of assembly for high-volume tape-and-reel requirements.

#### **Ordering Information**

NLB-300

NLB-300-T1

NLB-300-E

NLB-300-E

NBB-X-K1

Cascadable Broadband GaAs MMIC Amplifier DC to 10GHz

Tape & Reel, 1000 Pieces

Fully Assembled Evaluation Board

Extended Frequency InGaP Amp Designer's Tool Kit

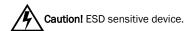
☐ GaAs HBT	☐ SiGe BiCMOS	☐ GaAs pHEMT	☐ GaN HEMT
☐_GaAs MESFET	☐ Si BiCMOS	☐ Si CMOS	
☐ GaAs MESFET ✓ InGaP HBT	☐ SiGe HBT	☐ Si BJT	



### **Absolute Maximum Ratings**

Parameter	Rating	Unit
RF Input Power	+20	dBm
Power Dissipation	300	mW
Device Current	70	mA
Channel Temperature	200	°C
Operating Temperature	-45 to +85	°C
Storage Temperature	-65 to +150	°C

Exceeding any one or a combination of these limits may cause permanent damage.



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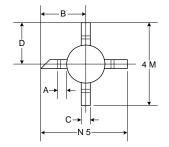
RoHS status based on EUDirective 2002/95/EC (at time of this document revision)

Davamatav	Specification			Unit	Condition	
Parameter	Min.	Тур.	Max.	Unit	Condition	
Overall					$V_D$ =+3.8V, $I_{CC}$ =50 mA, $Z_0$ =50 $\Omega$ , $T_A$ =+25 °C	
Small Signal Power Gain, S21	12.0	13.0		dB	f=0.1GHz to 1.0GHz	
		10.7		dB	f=1.0GHz to 4.0GHz	
		8.9		dB	f=4.0GHz to 6.0GHz	
	8.5	8.9		dB	f=6.0GHz to 10.0GHz	
		8.5		dB	f=10.0GHz to 12.0GHz	
Gain Flatness, GF		±0.1		dB	f=5.0GHz to 10.0GHz	
Input VSWR		2.2:1			f=0.1GHz to 4.0GHz	
		2.8:1			f=4.0GHz to 7.0GHz	
		2.0:1			f=7.0GHz to 12.0GHz	
Output VSWR		2.2:1			f=0.1GHz to 4.0GHz	
		2.9:1			f=4.0GHz to 7.0GHz	
		2.4:1			f=7.0GHz to 12.0GHz	
Output Power @ -1dB Compression, P1dB		11.1		dBm	f=2.0GHz	
		14.1		dBm	f=6.0GHz	
		12.7		dBm	f=10.0GHz	
Noise Figure, NF		4.9		dB	f=3.0GHz	
Third Order Intercept, IP3		+28.6		dBm	f=2.0GHz	
		+27.0			f=6.0GHz	
Reverse Isolation, S12		-16		dB	f=0.1GHz to 20.0GHz	
Device Voltage, V <sub>D</sub>	3.6	3.8	4.2	V		
Gain Temperature Coefficient, $\delta G_T/\delta T$		-0.0015		dB/°C		
MTTF versus Temperature @ I <sub>CC</sub> =50 mA						
		OE.		°C		
Case Temperature Junction Temperature		85 113		°C		
MTTF		>1,000,000		-		
Thermal Resistance		>1,000,000		hours		
		4.47		90.04		
$\theta_{ extsf{JC}}$		147		°C/W	$\frac{J_T - T_{CASE}}{V_D \cdot I_{CC}} = \theta_{JC}(^{\circ}C/Watt)$	

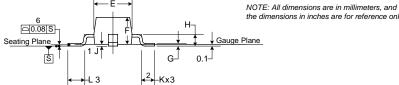


Pin	Function	Description	Interface Schematic
1	RF IN	RF input pin. This pin is NOT internally DC-blocked. A DC-blocking capacitor, suitable for the frequency of operation, should be used in most applications. DC coupling of the input is not allowed, because this will override the internal feedback loop and cause temperature instability.	
2	GND	Ground connection. For best performance, keep traces physically short and connect immediately to ground plane.	
3	RF OUT	RF output and bias pin. Biasing is accomplished with an external series resistor and choke inductor to $V_{CC}$ . The resistor is selected to set the DC current into this pin to a desired level. The resistor value is determined by the following equation: $R = \frac{(V_{CC} - V_{DEVICE})}{I_{CC}}$ Care should also be taken in the resistor selection to ensure that the current into the part never exceeds maximum datasheet operating current over the planned operating temperature. This means that a resistor between the supply and this pin is always required, even if a supply near 5.0V is available, to provide DC feedback to prevent thermal runaway. Because DC is present on this pin, a DC-blocking capacitor, suitable for the frequency of operation, should be used in most applications. The supply side of the bias network should also be well bypassed.	RF IN O
4	GND	Same as pin 2.	

# **Package Drawing**

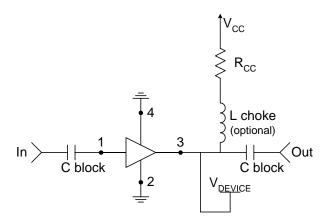


g	MIL	LIMET	ERS	INCHES			
Syn	Min. Nom. Max.		Min.	Max.			
Α	0.535 REF.			0.	021 RE	F.	
В	2.39	2.54	2.69	0.094	0.100	0.106	
С	0.436	0.510	0.586	0.017	0.020	0.023	
D	2.19	2.34	2.49	0.086	0.092	0.098	
Е	1.91	2.16	2.41	0.075	0.085	0.095	
F	1.32	1.52	1.72	0.052	0.060	0.068	
G	0.10	0.15 0.20		0.004	0.006	0.008	
Н	0.535	0.660	0.785	0.021	0.026	0.031	
J	0.05	0.10	0.15	0.002	0.004	0.006	
K	0.65	0.75	0.85	0.025	0.029	0.033	
L	0.85	0.95	1.05	0.033	0.037	0.041	
М	4.53	4.68	4.83	0.178	0.184	0.190	
N	4.73	4.88	5.03	0.186	0.192	0.198	
	BCDEFGHJKLM	A 0. B 2.39 C 0.436 D 2.19 E 1.91 F 1.32 G 0.10 H 0.535 J 0.05 K 0.65 L 0.85 M 4.53	A 0.535 RE B 2.39 2.54 C 0.436 0.510 D 2.19 2.34 E 1.91 2.16 F 1.32 1.52 G 0.10 0.15 H 0.535 0.660 J 0.05 0.10 K 0.65 0.75 L 0.85 0.95 M 4.53 4.68	A 0.535 REF.  B 2.39 2.54 2.69 C 0.436 0.510 0.586 D 2.19 2.34 2.49 E 1.91 2.16 2.41 F 1.32 1.52 1.72 G 0.10 0.15 0.20 H 0.535 0.660 0.785 J 0.05 0.10 0.15 K 0.65 0.75 0.85 L 0.85 0.95 1.05 M 4.53 4.68 4.83	A         0.535 REF.         0.           B         2.39         2.54         2.69         0.094           C         0.436         0.510         0.586         0.017           D         2.19         2.34         2.49         0.086           E         1.91         2.16         2.41         0.075           F         1.32         1.52         1.72         0.052           G         0.10         0.15         0.20         0.004           H         0.535         0.660         0.785         0.021           J         0.05         0.10         0.15         0.02           L         0.85         0.95         1.05         0.033           M         4.53         4.68         4.83         0.178	A         0.535 REF.         0.021 RE           B         2.39         2.54         2.69         0.094         0.100           C         0.436         0.510         0.586         0.017         0.020           D         2.19         2.34         2.49         0.086         0.092           E         1.32         1.52         1.72         0.052         0.060           G         0.10         0.15         0.20         0.004         0.006           H         0.535         0.660         0.785         0.021         0.026           J         0.05         0.10         0.15         0.020         0.004           K         0.65         0.75         0.85         0.025         0.029           L         0.85         0.95         1.05         0.033         0.037           M         4.53         4.68         4.83         0.178         0.184	



## **Typical Bias Configuration**

Application notes related to biasing circuit, device footprint, and thermal considerations are available on request.



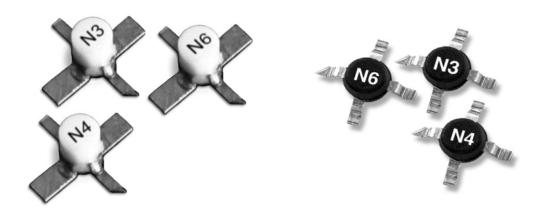
Recommended Bias Resistor Values						
Supply Voltage, V <sub>CC</sub> (V)         5         8         10         12         15         20						
Bias Resistor, $R_{CC}(\Omega)$	22	82	122	162	222	322



# Extended Frequency InGaP Amplifier Designer's Tool Kit NBB-X-K1

This tool kit was created to assist in the design-in of the RFMD NBB- and NLB-series InGap HBT gain block amplifiers. Each tool kit contains the following.

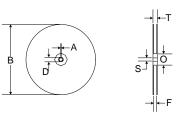
- 5 each NBB-300, NBB-310 and NBB-400 Ceramic Micro-X Amplifiers
- 5 each NLB-300, NLB-310 and NLB-400 Plastic Micro-X Amplifiers
- 2 Broadband Evaluation Boards and High Frequency SMA Connectors
- Broadband Bias Instructions and Specification Summary Index for ease of operation



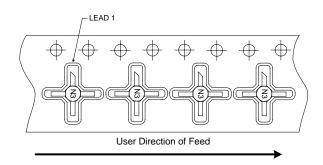


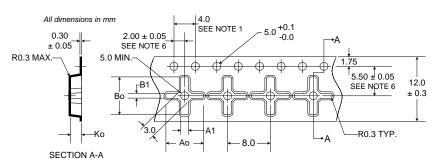
## **Tape and Reel Dimensions**

All Dimensions in Millimeters



14.732 mm (7") REEL			Plastic, Micro-X		
	ITEMS	SYMBOL	SIZE (mm)	SIZE (inches)	
	Diameter	В	178 +0.25/-4.0	7.0 +0.079/-0.158	
FLANGE	Thickness	Т	18.4 MAX	0.724 MAX	
	Space Between Flange	F	12.8 +2.0	0.50 +0.08	
	Outer Diameter	0	76.2 REF	3.0 REF	
HUB	Spindle Hole Diameter	S	13.716 +0.5/-0.2	0.540 +0.020/-0.008	
пов	Key Slit Width	Α	1.5 MIN	0.059 MIN	
	Key Slit Diameter	D	20.2 MIN	0.795 MIN	

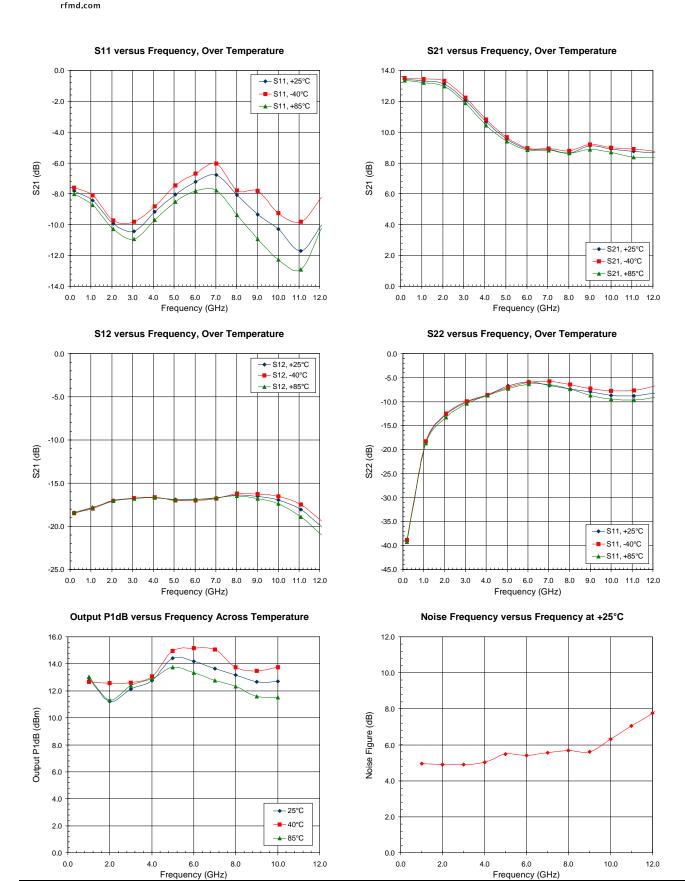




#### NOTES:

- 1. 10 sprocket hole pitch cumulative tolerance ±0.2.
- Camber not to exceed 1 mm in 100 mm.
   Material: PS+C.
- 4. Ao and Bo measured on a plane 0.3 mm above the bottom of the pocket.5. Ko measured from a plane on the inside bottom of the pocket to the surface of the carrier.
- 6. Pocket position relative to sprocket hole measured as true position of pocket, not pocket hole.
- Ao = 7.0 MM
- A1 = 1.8 MM
- Bo = 7.0 MM
- B1 = 1.3 MM
- Ko = 2.1 MM







Note: The s-parameter gain results shown include device performance as well as evaluation board and connector loss variations. The insertion losses of the evaluation board and connectors are as follows:

1GHz to 4GHz=-0.06dB 5GHz to 9GHz=-0.22dB 10GHz to 14GHz=-0.50dB 15GHz to 20GHz=-1.08dB

### **RoHS\* Banned Material Content**

Bill of Materials	Parts Per Million (PPM)							
Dili di Materiais	Pb	Cd	Hg	Cr VI	PBB	PBDE		
Die	0	0	0	0	0	0		
Molding Compound	0	0	0	0	0	0		
Lead Frame	0	0	0	0	0	0		
Die Attach Epoxy	0	0	0	0	0	0		
Wire	0	0	0	0	0	0		
Solder Plating	0	0	0	0	0	0		

This RoHS banned material content declaration was prepared solely on information, including analytical data, provided to RFMD by its suppliers, and applies to the Bill of Materials (BOM) revision noted

<sup>\*</sup> DIRECTIVE 2002/95/EC OF THE EUROPEAN PARLIAMENT AND OF THE COUNCIL of 27 January 2003 on the restriction of the use of certain hazardous substances in electrical and electronic equipment